

SINGLE STAGE

(FLYBACK-DISCONTINUOUS)

POWER SUPPLY

MAINTENANCE MANUAL

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SINGLE STAGE (FLYBACK-DISCONTINUOUS) POWER SUPPLY TABLE OF CONTENTS

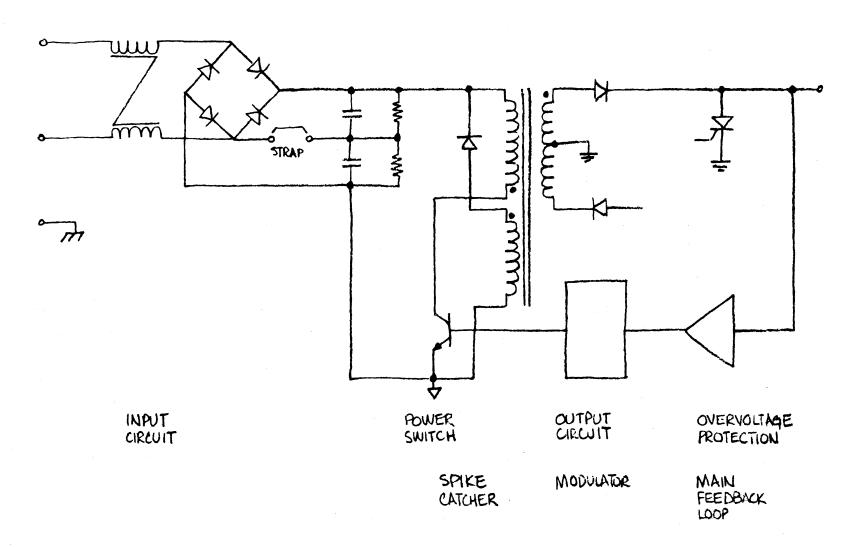
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NOTE: The descriptions and component designations in this manual refer to the OL25-1001. However, all flyback-discontinuous supplies have analogous circuits which operate in a manner similar to the OL25-1001.

FIGURE 1

SINGLE STAGE (FLYBACK-DISCONTINUOUS) POWER SUPPLY

SIMPLIFIED SCHEMATIC



SINGLE STAGE (FLYBACK-DISCONTINUOUS) POWER SUPPLY

GENERAL THEORY OF OPERATION

The flyback converter-discontinuous mode derives its name from the fact that during each cycle a point is reached where current ceases to flow in either the primary or secondary windings. Thus, the flux in the transformer virtually "ceases" at some point in the cycle.

The converter design is extremely simple. This class of converter operates as a blocking oscillator under the following control law: The output is linearly proportional to the current flowing in the power switch when it turns off, i.e., the output voltage is proportional to the amount of energy loaded into the core in the form of a magnetic field. The greater the field, the higher the output voltage.

The feedback loop regulates output voltages by adjusting the turn-off point of the power switch. Switching frequency is around 25 kHz at full power, and operating frequency is inversely proportional to the output power. The duty cycle remains relatively constant for a constant line voltage.

SINGLE STAGE (FLYBACK-DISCONTINUOUS) POWER SUPPLY

DETAILED THEORY OF OPERATION

Input Circuit

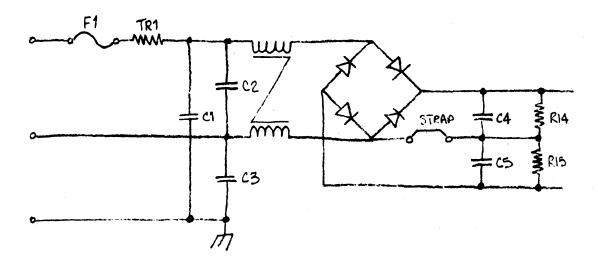


Figure 2
INPUT CIRCUIT

- a. <u>RFI Filter</u>. Consists of L1, C1, C2 and C3. The purpose of this circuit is to filter out 20 kHz and above switching noise, preventing it from being transmitted back out the input line.
- b. <u>Fuse</u>. Fl is included to protect the PC board traces and to reduce fire and personal hazard in the event of catastrophic supply failure. Boschert strongly recommends the use of an additional external line fuse for further protection. A Littlefuse 3AG series or equivalent rated at 1-1/2 A is sufficient.

c. <u>Voltage Doubler Circuit</u> is a method of allowing the supply to operate from either a 115 VAC or 220 VAC source. In the 115 VAC mode (strap in) capacitors C4 and C5 charge on alternate line half cycles. The voltage which appears across each capacitor is approximately the peak input voltage (about 150 V). The total voltage to the switching regulator is about 300 V.

In the 220 VAC mode (strap out) the input rectifiers act as a full wave bridge rectifier, charging C4 and C5 to approximately the peak input voltage (about 300 V). Resistors R14 and R15 force C4 and C5 to charge equally, and discharge the capacitors when the power is turned off.

- d. <u>Inrush Limiting</u> is accomplished with thermistor R1. When cool, its resistance is high. When the supply is initially turned on, it prevents a huge surge current from flowing into C4 and C5, which are initially discharged. In operation, the normal input current quickly heats R1 decreasing its resistance by a factor of about 10. Since C4 and C5 are fully charged by this time, there is no further need for surge limiting. When the supply is turned off, R1 cools and C4 and C5 discharge. The circuit is designed so that the discharge time constant and cooling time constant are roughly equal. Thus, if the supply is turned back on before R1 has fully cooled, the inrush current is limited by a combination of the charge remaining on C4 and C5, and the resistance of R1.
- e. <u>Energy Storage</u> in capacitors C4 and C5 insure that the supply will continue to operate within regulation limits for a minimum of 16 mS after the input line has fallen below limits or failed. This allows time for information in a volatile memory to be stored permanently before system failure following line interruption or line brownout.

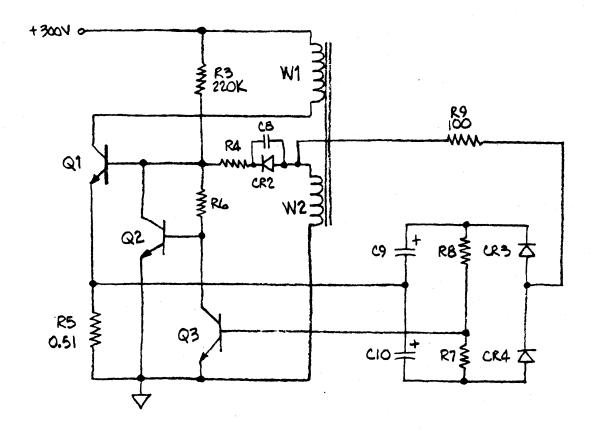


Figure 3
POWER SWITCH CIRCUIT

In the first half of the power cycle, switch Ql will turn on and load energy into the transformer in the form of a magnetic field. Switch Ql will then turn off.

The second half of the power cycle is described in the Output Circuit description.

When power (+300) is applied, a small trickle current through R3 begins to bias Q1 on. As the full supply voltage +Vcc begins to appear across winding one (W1), transformer action induces about 6 V across winding two (W2). CR2 is forward biased, and regenerative feedback current from W2 quickly forces Q1 into saturation.

Now the current in W1 begins to ramp up linearly. A voltage ramp also appears across R5 due to the current ramp.

Meanwhile, W2 is also charging C9 to about +6 V via R9 and CR3. When this voltage rises high enough, Q3 is biased on. This also turns on current buffer Q2. Q1 begins to turn off because Q2 and Q3 are conducting base drive away from Q1. The current ramp in W1 stops rising because Q1 is beta limited and is coming out of saturation. The rate of current change (di/dt) has become negative, therefore the voltage polarity on the windings must change. W2 has now reversed voltage (the dotted end is now positive) and it helps turn off Q1 by completely backbiasing the base emitter junction via R5, R4 and C8. When Q1 is off, and the voltage across all transformer windings has reversed, C10 is charged via CR4 and R9.

(The second half of the cycle is continued in the Output Circuit Section.)

Output Circuit

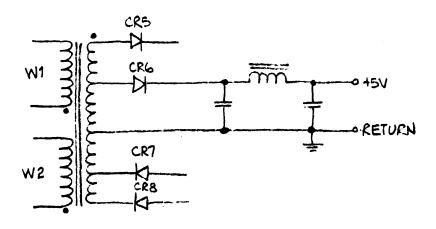


Figure 4
OUTPUT CIRCUIT

This section describes the second half of the power cycle. The first half was described in the Power Switch Circuit description.

In the first half of the power cycle, switch Q1 turned on and loaded energy into the transformer in the form of a magnetic field. Q1 then turned off, and the collapsing magnetic field is now driving positive the dotted ends of all transformer windings, looking for a discharge path.

As the voltage rises on the dotted ends of all the windings, it will eventually forward bias diodes CR5 through CR8. The magnetic field in the transformer rapidly collapses by discharging energy into the outputs. The C-L-C PI filters in the output store and filter this energy.

When the magnetic field has sufficiently collapsed, diodes CR5 through CR8 stop conducting. There is enough energy left in the core to allow it to "ring back." That is, the current in the secondary suddenly reduces its rate of discharge when the diodes stop conducting. di/dt changes, and the dotted end of all windings becomes negative. QI begins to conduct again. This is the start of the next cycle.

Modulator Circuit

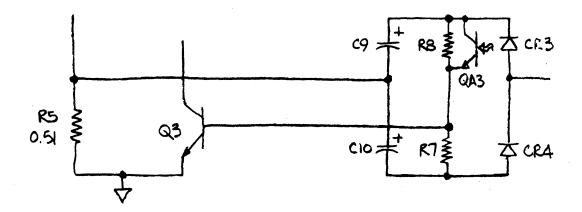


Figure 5
MODULATOR CIRCUIT

The modulator varies the point at which Q1 turns off thereby regulating the output voltages.

In the absence of any feedback loop, Q1 would continue loading energy into the core until a fixed point was reached. This fixed amount of energy would be subsequently discharged into the load. If the load changed, the output voltage would change because Q1 was still loading a fixed amount of energy into the transformer. To compensate for load changes, the feedback loop varies the point at which Q1 turns off, thereby varying the amount of energy loaded in the core. The greater the load, the more energy is loaded into the core. This accomplishes regulation.

During normal operation, C9 and C10 are each charged to about 6 V, or about 12 V total. Since the junction of C9 and C10 is referenced essentially to ground via R5, the base-emitter junction of Q3 must be back

biased due to the base voltage determined by dividers R8 and R9. When Q1 turns on, the voltage across R5 ramps up. This makes the entire modulator circuit voltage rise relative to ground. Eventually a point is reached where Q3 is biased on, which turns off Q1, terminating the first half of the cycle.

The point at which Q3 turns on can be varied by changing the ratio of R8 to R9. The opto-isolator transistor, Q A3, acts as a variable resistor to change this ratio. If the +5 V output tended to fall due to an increase in load, for example, the opto-isolator transistor would turn off, making the base voltage on Q3 more negative. It would take longer for the voltage ramp on R5 to reach a point where Q3 was biased on. More energy would be loaded into the transformer, which would be discharged into the load, raising the output voltage.

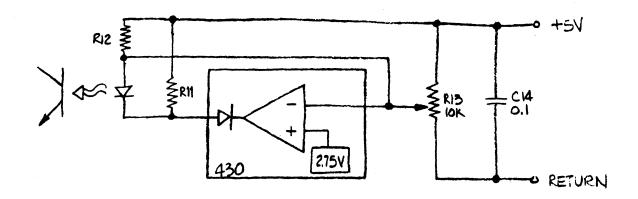


Figure 6
MAIN FEEDBACK LOOP

This circuit maintains the +5 V output at a constant voltage. The loop compares a voltage divided down from the +5 V output via R13 to a 2.75 V reference within the 430 integrated circuit. A proportional current is sunk by the 430, driving the opto-isolator diode via R12. R11 provides bias current for the 430. C13 frequency compensates the loop.

The auxiliary (other than +5V) outputs are "semi-regulated." That is, because of the tight magnetic coupling of all the outputs, an increased load on an auxiliary output lowers the volts per turn of the transformer, which is reflected in the +5V output (which begins to go down). The feedback loop works to restore the +5V output, and in so doing partially compensates for the load change on the auxiliary output. Boschert calls this "semi-regulation."

Overvoltage Protection

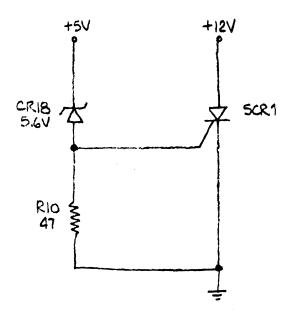


Figure 7
OVP CIRCUIT

The purpose of overvoltage protection is to protect the user's load from an overvoltage condition caused by supply failure. All standard Boschert supplies incorporate this feature. Also, Boschert supplies require OVP on only one output. If all outputs have simple passive filtering, and if one output should rise due to supply failure, they all would rise. Thus, protection is needed on only one output.

If the +5 V output exceeds 6.25 V ± 0.75 V, the SCR pulls the +12 V supply down (to about 1.5 V) and activates the short circuit protection. The +12 V supply is used to insure clean foldback. Pulling the +5 V down to 1.5 V might not exceed the foldback point, and might simply deliver a lot of power to the SCR, destroying it.

This circuit can be cleared by cycling line power.

Spike Catcher Circuit

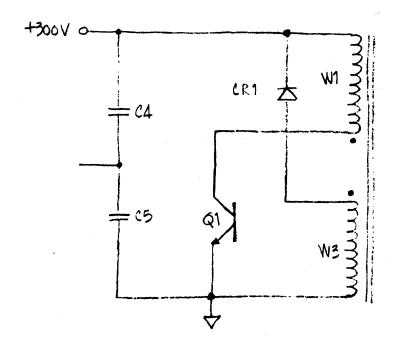


Figure 8
SPIKE CATCHER CIRCUIT

The purpose of the spike catcher circuit is to prevent high voltage turn off transients on the collector of Q1 from destroying it.

C4 and C5 are the input capacitors, charged to about 300 V. When Q1 has finished loading energy into the transformer core and turns off, W1 begins to look like a current source. Its dotted end becomes positive and the voltage rises. Although the discharge path is ultimately through the output, there is enough leakage inductance in W1 to allow the voltage to rise to a destructive value. However, as the dotted end of W1 reaches 300 V, so does the dotted end of W3. (They both have the same number of turns.) As the voltage tries to rise further, CR1 conducts and clamps the collector voltage to 600 V. This prevents the destruction of Q1.

SINGLE STAGE (FLYBACK-DISCONTINUOUS) POWER SUPPLY

TROUBLESHOOTING INSTRUCTIONS

EQUIPMENT NEEDED:

- 1 50 MHz oscilloscope with isolated ground
- 1 DVM or VTVM with isolated ground
- 1 Isolation transformer
- 1 1A Variac (0-130Vac)
- 2 Adjustable lab supplies, 0-20VDC @ 500mA, with adjustable current limit. A good supply of resistors and clip leads.
- 1 AC voltmeter (0-130Vac)

CAUTION: Lethal voltages are present in this supply. Only authorized service technicians should perform these tests. Use AC isolated equipment in all tests.

NOTE: It is desirable to use an AC isolation transformer in performing all tests to minimize shock hazard. The kVA rating of this transformer should be three times the maximum supply power to avoid AC line waveform distortion.

NOTE: Steps 1 thru 9 are intended as passive tests to be performed with the supply completely disconnected from line power.

To prevent any oscillations, connect jumper wire from the anode of CR5 to the cathode of CR8.

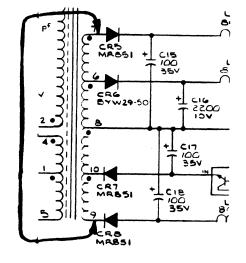


Figure 9

- Use DVM to check the continuity of the input fuse (F1), thermistor (TR1) bridge rectifier (BR-1), diodes CR1 thru CR8 for proper continuity. Replace as necessary.
- 2. Connect a lab supply (15V with current limit at 500mA) across the AC input terminals. After an initial surge, no current should flow. Reverse the leads and repeat. If any current flows, check the input section for shorts.

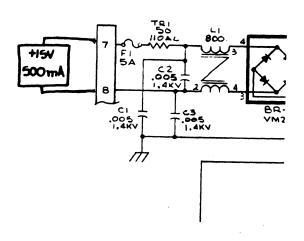


Figure 10

3. Check the operation of Q1. Set the lab supply for 15V with the current limit set for 500mA. Connect a 220 ohm resistor in series with the positive output and connect this to the 115V input terminals as shown. With the DVM across the collector-emitter terminals of Q1, momentarily parallel R3 (220K) with a 100 ohm 1/4W resistor. Is voltage drop across Q1 as follows?

Vce (Q1)

R3 not paralleled 13.5V ± 1 V

R3 paralleled 1.5V $\pm 1V$

Yes - Go to Step 4

No - Replace Q1 and repeat Step 3

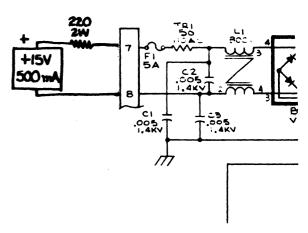


Figure 11

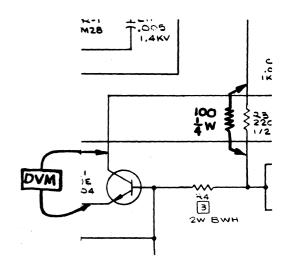


Figure 12

4. Check the operation of Q2 and Q3. Use a DVM or a lab supply with the output set at +5V, current limit at 10mA. Check the continuity of Q2 and Q3 for opens or shorts. Replace as necessary. Then go to Step 5.

5. Check overvoltage protection operation.

Set both supplies to +5.0V with the current limit set to 100mA. Connect the first supply across the +5V output and return. Connect the second supply across the +12V output and return. (The +12V output should be the one which the anode of the SCR is connected to. If not, the supply should be connected to the output which is connected to the SCR anode. Now slowly ramp up the voltage on the first supply. At 6.25V ±0.75V on the first supply, does the second supply suddenly go into current limit?

6. Check CR9. Set lab supply current limit to 100mA and the voltage to 0. Connect the positive terminal to the +5V output and the negative terminal to the return. Put a DVM across CR9. Turn the voltage slowly up to 10V, or to the point where the supply limits. Does the DVM voltage ramp up to 5.6V + 0.25V and stop?

Yes - Replace CR4, and Repeat Step 5 No - Replace SCR-1 and Repeat Step 5

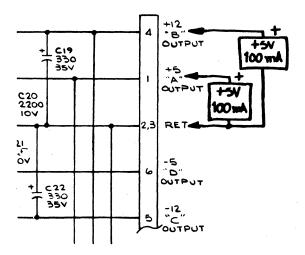


Figure 13

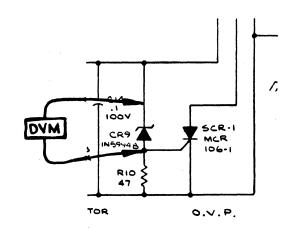


Figure 14

7. Check A2. Set lab supply to +5.0V (Ilim = 300 mA) across +5V output. Read voltage across R11 with DVM while adjusting pot R13. Are the voltages as follows?

R13 fully clockwise $3.0V \pm 0.5V$ R13 fully counterclockwise $0.2V \pm 0.2V$

Yes - Go to Step 8
No - Replace A2 and repeat Step 7

8. Check A3 and modulator. Hook up first supply (+5V, Ilim = 300mA) across the +5V output. Hook up second supply (+5V, Ilim = 300mA) across the modulator. Watch the voltage across R8 with the DVM while adjusting R13. Are the voltages as follows?

R13 fully clockwise $3.0V \pm 0.5V$ R13 fully counterclockwise $0.3V \pm 0.2V$

Yes - Go to Step 9 No - Replace A3, repeat Step 8

NOTE: This concludes all the passive tests. Disconnect all supplies, DVM, and jumper from CR5 to CR8 before proceeding.

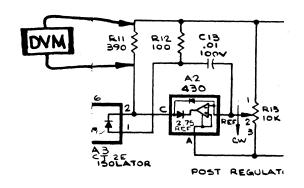


Figure 15

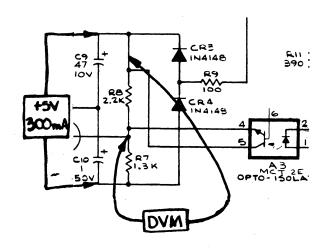


Figure 16

9. Disconnect all connections made for the passive tests. Plug the supply into the Variac and turn the voltage up to 115Vac. Is the voltage across the input capacitor(s) within the following limits?

Supply designed for 115Vac only 150V ±10V

Supply designed for 220Vac only $300V \pm 20V$

Supply 115/220 strappable $300V \pm 20V$

Yes - Go to Step 10

No - Look for faulty component in input circuit. Repeat Step 9.

10. Use an oscilloscope to check the voltage waveform on the collector of Ql with respect to emitter.

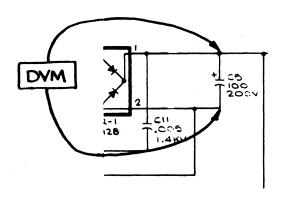


Figure 17

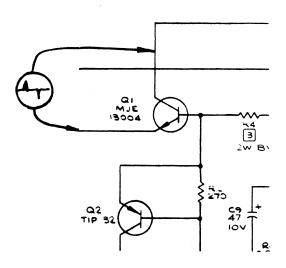


Figure 18

With proper waveform, notice that duty cycle is roughly 50%.

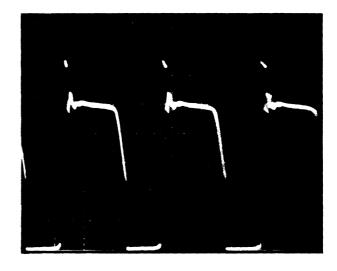


Figure 19
Proper Waveform
Horizontal - 5 µs/Div.
Vertical - 50V/Div.

With no output load, the supply "burps" every 120 ms or so. This is known as the burp mode. (The supply waits about 120ms, tries to turn on, the SCR fires because there is no +5V load to keep it from overshooting, and it folds back and waits 120ms again.)

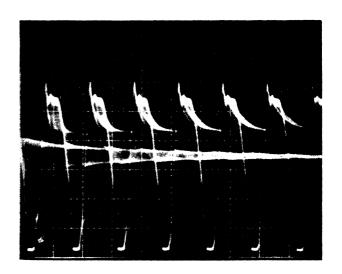


Figure 20 No load waveform Horzontal - 5µs/Div. Vertical - 50V/Div.

In an overload condition, all output voltages and currents are very low.

Notice the very short duty cyle (about 12%) which typifies the overload condition.

Under a heavy overload, the supply "burps" every 500ms or so.

With an output short, the supply goes into its "burp" mode again. The period is about 160ms between "burps."

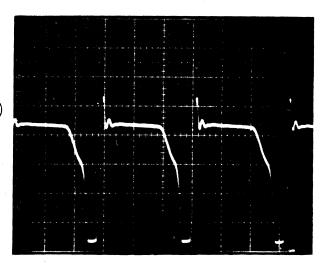


Figure 21 Overload Waveform Horizontal 5µs/Div. Vertical 50V/Div.

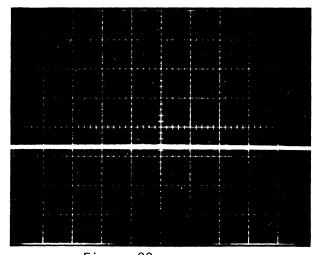


Figure 22 Output Short Waveform Horizontal 5µs/Div. Vertical 50V/Div.

11. Use the oscilloscope on each or any output to determine if any faults are occurring.

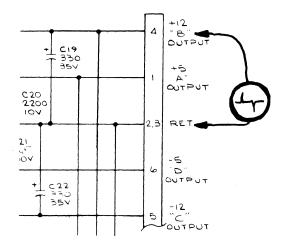


Figure 23

+12V Output Proper Waveform

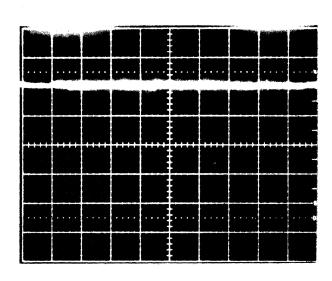


Figure 24
Proper Output Waveform
Horizontal - 50ms/Div.
Vertical - 2V/Div.

The sharp rise and fall time of an output is due to unloaded outputs. Generally only an SCR firing can cause such a rapid fall time on an output.

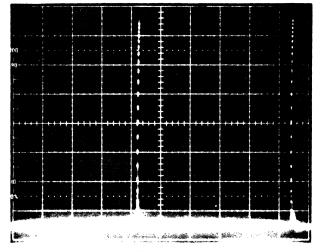


Figure 25
No Load on Outputs
Horizontal - 50ms/Div.
Vertical - 2V/Div.

The much longer fall time of an output indicates a short on some other output than the one measured.

No rise or fall indicates the short is on this output.

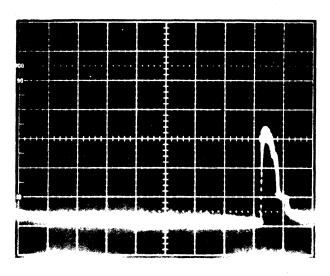


Figure 26

Short on some other output

Horizontal - 50ms/Div.

Vertical - 2V/Div.

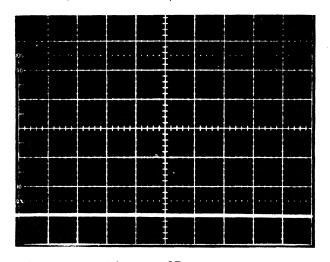


Figure 27
Short on this output
Horizontal - 50ms/Div.
Vertical - 2V/Div.

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FACTORY SELECT PROCEDURE FOR R4

- 1. Q. SET LOAD AT : 3A, +5V .5A, +12V .5A, -12V .5A, -5V
 - b. SET LINE VOLTAGE AT GOVAC C. SELECT RA FOR +5V OUTPUT TO BE A.95V & +5V OUTPUT & 5.05V
- 2. Q. SET LINE VOLTAGE TO HOVAC b. SET ALL LOADS TO O AMP C. LOAD SV SLOWLY SV OUTPUT VOLTAGE MUST REGULATE TO SV ± 0.050V AT +5V OUTPUT CURRENT OF \$ 0.8 AMP

CURRENT MAX LIMIT TEST

- a. SET LINE VOLTAGE TO 110 V AC
- b. SET LOAD TO: +5V 3A ±12V 0.5A -5V 0.5A
- C. INCREASE OUTPUT CURRENT OF +5V

 AND NOTE MAXIMUM CURRENT.

 MAXIMUM CURRENT, I MAX, MUST

 BE: 6.5 A \leq I MAX \leq 8.0 A. IF CURRENT

 LIMIT IS NOT IN SPEC, CHANGE RIH TO ANDTHER

 VALUE BETWEEN 470 AND IDK (NOMINAL VALUE

 INSTALLED IS 1.2 K). LOWER THE RESISTOR

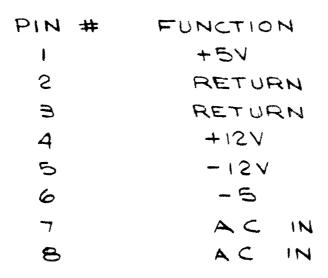
 VALUE TO BRING THE CURRENT LIMIT DOWN

 OR RAISE THE VALUE TO INCREASE THE CURRENT

 LIMIT.

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.XXX ± - ± -	DRAWNERANK KRALOVETZ	7-13	E	INO	1) T	CH	AR"	F	
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FINISH	a facilism	9/19/18	SIZE A	CODE IDEN	IT NO.	DRAWING NO) 26	7	REV
DO NOT SCALE DRAWING			SCALE	N/A		A. N. 1944 (1974) (1974) (1974) (1974) (1974) (1974) (1974) (1974) (1974) (1974) (1974) (1974) (1974) (1974)	SHEET	OF	1
A BISHOP GRAPHICS/ACCHERES					2 -	NOV - 1300			

BISHOP GRAPHICS/ACCUPRESS REORDER NO. A-7825 20267 RKS

		REVISIONS		
ECO	LTR	DESCRIPTION	DATE	APPROVED
1986	4	PRE-RELEASED FOR PRODUCTION	5-2-78	Fin
<i>22</i> 59	A	RELEASED FOR PRODUCTION	9-13-78	8771
2369	В	ITEM 28 WAS PIN 2061 (NO DESCRIPTION CHG.);	12-13-78	87. X.
		ITEM 45 WAS PIN 1140 (NO DESCRIPTION CHG); ADDED		WSL
		P/N 3810 TO ITEM 86 22 L ZW BWH SAT R4.		
2413	C	ASSY CHG ONLY	12-13-78	NSA NSA
2623	D	ITEM 76 WAS 1.3K (NOT F.S) ADDED R7 REF TO NOTE 8. ADDITIONAL CHGS NOT PER ECO AS FOLLOWS: DELETED NOTE B REF AT ITEM 86 & NOTE 4 REF AT ITEMS 107 & 108		liş
27 99	E	ITEM 76 WAS 1.3K; DELETED ITEM 8 P/N 7641, 45 P/N 10-10056-01, 59 P/N 1081; 61 P/N 142; ADDED ITEMS 4,5,6.	9-25-19	51. W
33 35	F	CHG'D NOTE B , ITEM 16 WAS P/N 3075, ITEM 81 TVA PENDING WAS P/N 3080 & ADDED ITEM 77	5-1-79	RKS (MWG
				Dell VAA
3915	Н	INCORPORATED	3-5-80	Rest XIO

All drawings supplied for the BOSCHERT power supply are the property of BOSCHERT INCORPORATED and shall not be reproduced or copied or used in whole or in part as the basis for manufacture or sale of the items shown therein without the express written permission of an officer of BOSCHERT INCORPORATED.

SEE SEPARATE ASSY, DWG NO. 40006 "C" SIZE

		UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCES ARE:	CONTRACT NO.					nachar!			
		FRACTIONS DECIMALS ANGLES	APPROVALS	DATE					<u> </u>		
		.xxx ± — ± — .xxx ± —	DRAWN FRANK KRALOVETZ	4-5	ASS	SEMB	LY.	POWE	R SU	PPL.	~
		MATERIAL	CHECKER	52-18	1	WATT	•	115V	, ,		
			WStee	9-19-78							
FINAL	25-1001	FINISH	all Colon End	4/14/28	SIZE	CODE IDEN	T NO.	45-4		16	REV
NEXT ASSY	USED ON		a Jackson	8/29/18	^			754			<u>)</u>
APPLIC	CATION	DO NOT SCALE DRAWING		•	SCALE	N/A			SHEET	OF (6

ìREVISIONS									
LTR	DESCRIPTION	DATE	APPROVED						
1		Į.	t						

8 FACTORY SELECT R4 & R14

- 7 FOR QA SPEC SEE DWG NO. 20270
- 6. FOR PINOUT CHART SEE DWG NO 20267
- 5, FOR OUTLINE INSTALLATION SEE DWG NO. 20168
- 4.
- 3. FOR TEST SPEC SEE DWG NO. 20269
- 2.
- 1. FOR SCHEMATIC SEE 20139

NOTES:

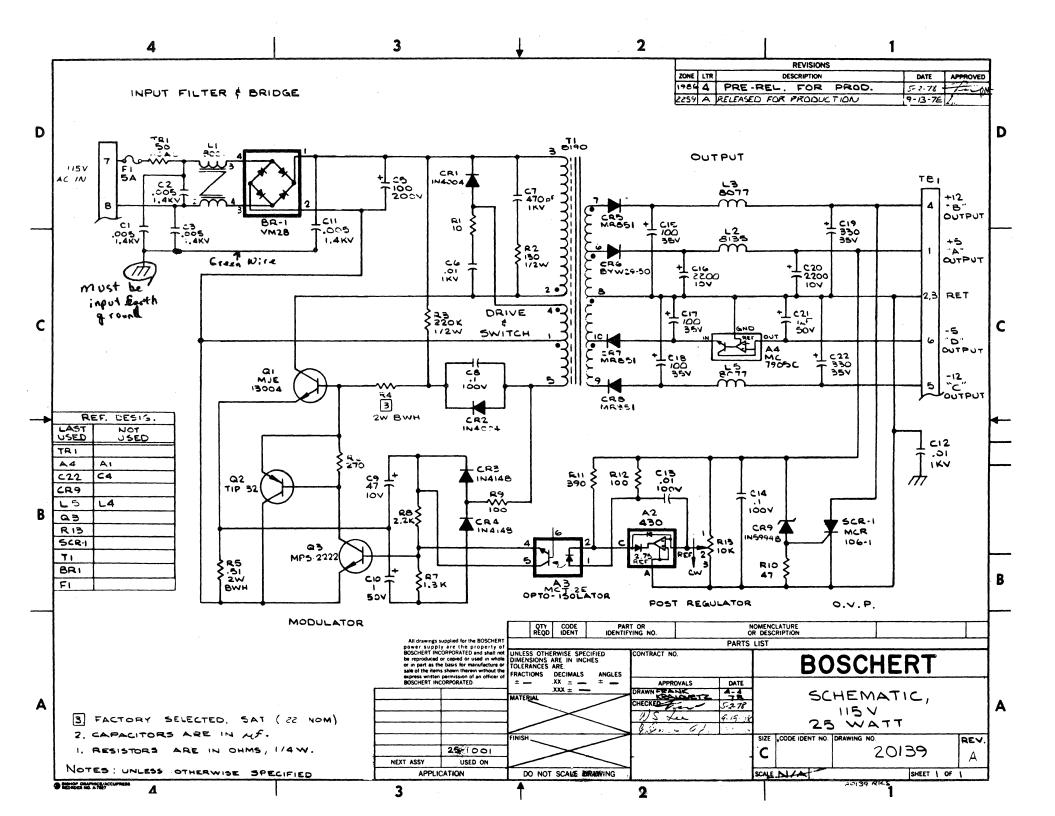
		UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCES ARE:	CONTRACT NO.				hanabar	Ł	
		FRACTIONS DECIMALS ANGLES	APPROVALS	DATE					
		± .XX ± ±XXX ±	DBAWN		ASSEMBLY, POWER SUPPL			01 >	
		MATERIAL	CHECKED		TASSEMBLY, POWER SUPPLY				
					2	.5W,	115 \		
		FINISH			SIZE	CODE IDENT NO			Re
NEXT ASSX	11050 011	Tition			A		4-5-4	10006	J
NEXT ASSY APPLICA	USED ON	DO NOT SCALE DRAWING			SCALE	NIA		SHEET 2 OF	6

TEM	ART NO.	QTY	SCRIPTION	F DES
	51-9148	١	BOARD, P.C.	
2				
3				
4	43-10568	-	SUB ASSY , HEATSINK , RECTIFIER , BYW 29-50	CR6 REF
5	43-10570	1	SUB ASSY, HEATSINK, TRANSISTOR, MJE-13004	QI REF
0	43-10575		SUBASSY, HEATSINK, VOLT REG. I.C., MC 7905CT	A4 REF
7				
8	٠			
9				
10		,		
11	80-8001	.	INDUCTOR,	LI
12	80-8135	1	INDUCTOR,	L2
13	80-8077	2	INDUCTOR,	L3,5
14				
15				
16	80-8190	١	TRANSFORMER, 25W	TI
17		•		
18				
19	22-2060	1	CAP, CERM, 470 pf, ± 20%, 1000V	C7
20	22-2002	4	, .005uf, ±20%, 1.4KV	C1, 2, 3, 11
21	22-2005	2	, .01 mf , ± 20% , 1000V	C6,12
22	22-2008	2	, , , , ± 20%, 100V	C 8,14
23				
24				
25				
26			<u> </u>	
27	22-2059	1	CAP, CERM, .014F, ±20%, 100V	C13
TITLE	POWER SUI 5V 25WA			006 J N/A SHEET 3 OF 6

I _T EM	ART NO.	, QTY	SCRIPTION	F DES
28	20-2089		CAP, ELECT, 100 mf, +75%-10%, 200V	C5
29	20-2083	2	, 3304F, +75%-10%, 35V	C19, ZZ
30	20-2106	١	, 47 mf, +75% -10%, 10V	C9
31				
32	20-2119	3	100 uf +50% -10%, 35V	C15,17,18
33	20-2047	2	, ELECT , 2200, +, +50% - 10%, 10V	C16,20
34				
35				
36	21-2076	2	CAP , TANT , 14f , ± 10% , 50V	C10,21
37				
38				
39	10.1038	2	DIODE, GEN PUR, IN4004	CR1,2
40				•
41				
42	10-1013	2	GEN PUR, IN4148	CR 3, 4
43				
44	and the second s			
45				
46				
47				
48	10-1088	3	, FAST RECV, MR851	CR5,7,8
49				
50				
51	11-1014	-	DIODE , ZENER , INS994 B	CR9
52				
53	1141	1	BRIDGE, VM28	BRI
54				<u> </u>
TITLE	POWER SU			0006 J /A SHEET 4 OF 6

ITEM	art no.	QTY	SCRIPTION	F DES
55	16-1132	1	SCR, MCR106-1	3CR-1
56				
57	15-1001		1.C, OPTO-ISOLATOR, MCT-2E	A 3
58	14-1071	1	1.C, VOLTAGE REG , 430	A2
59				
60				
61				
	13-1055		TRANS, NPN, AMPL, MPS-2222	QB
	12-1146	1	TRANS, PNP, TIP32	Q2
64				
65	,			
66	83-7011		FUSE, SUB-MINI, SA	FI
67				
68				
69	30-3024		RES, C.F., 10 , +5% 1/4W	RI
70	3040	1	47 ,	RIO
71	3048	2	, 100	R9,12
72				
73	3058	1	, 270	R6
74	3062	1	RES C.F. , 390 , ±5% , 1/4 W	RII
75				
76	30747	١	RES , C.F. 1.2K, ± 5°/4, 1/4W,	R7
77	30745	1	RES . C.F. SAT , (1.2K NOM) ±5% 1/4W	R14 8
78				
79				
80				
81	30-3082	1	RES, C.F. , 2.7K , ±5% , 1/4W	RB
TITLE	POWER SUPP 15V, 25 WAT		DRAWING NO. 45-40	0006 T 1/A SHEET 5 OF 6

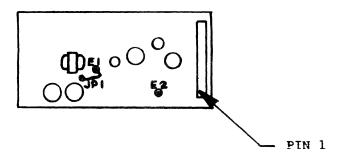
I _{TEM}	ART NO.	, QTY	SCRIPTION	RL DES
82	30-3251	1	RES, C.F., 130 , ±5%, 1/2W	R2
83				
84	30-3328	١	, C.F., 220K , ±5%, 1/2 W	RB
85				
86	31-3810	١	, C.C., SAT (ZZ & NOM.) 10%, BWH	R4 8
87				
88	35-3802	١	, W.W., .512 , ±10% , 2W BWH	RS
89	·			
90				
91				
92	36 - 3944	١	, POT, IOK	RIB
93				
94	39-3900	١	RES, SOL, THERMISTOR	TRI
95				
96	73-6058	1	TERMINAL STRIP	
97				
98				
99				
100	53-7742	١	LABEL, 115 VOLTS	
101	53-7740	1	LABEL, DANGER HIGH VOLTAGE	
102	53-7881	1	LABEL MODEL & SERIAL NUMBER	
103				
104		A/R	GLYPTAL, RED	
105	8 7- 7959		WIRE BARE ZZ AWG, .50LG	
106	87- 7960	1	WIRE, BARE, 22 AWG, -GOLG	
107	87-7464	ĺ	WIRE, GRN , 22 AWG , 3.00 x /4 x /4	
108				
TITLE			beechert Drawing no. 45-40	PEV REV
ı	POWER SUF		45-40	J006 I
11	5V; 25 WA	ユ T ア	SIZE A SCALE N	SHEET GOF G



APPLIC	CATION	REVISION						
NEXT ASSY	USED ON	REV	ECO		DES	CRIPTION	DATE	APPROVED
45-10157-01	XLZ5-1001	١,	_	PRE	REL	FOR PROD	11/6/20	le Sk
(5-1015Z-01 5-1015Z-0Z		Α		REL.	FOR	PROD.		Mo tolas
.5.5.52	7001		-					, , , , , , , , , , , , , , , , , , , ,

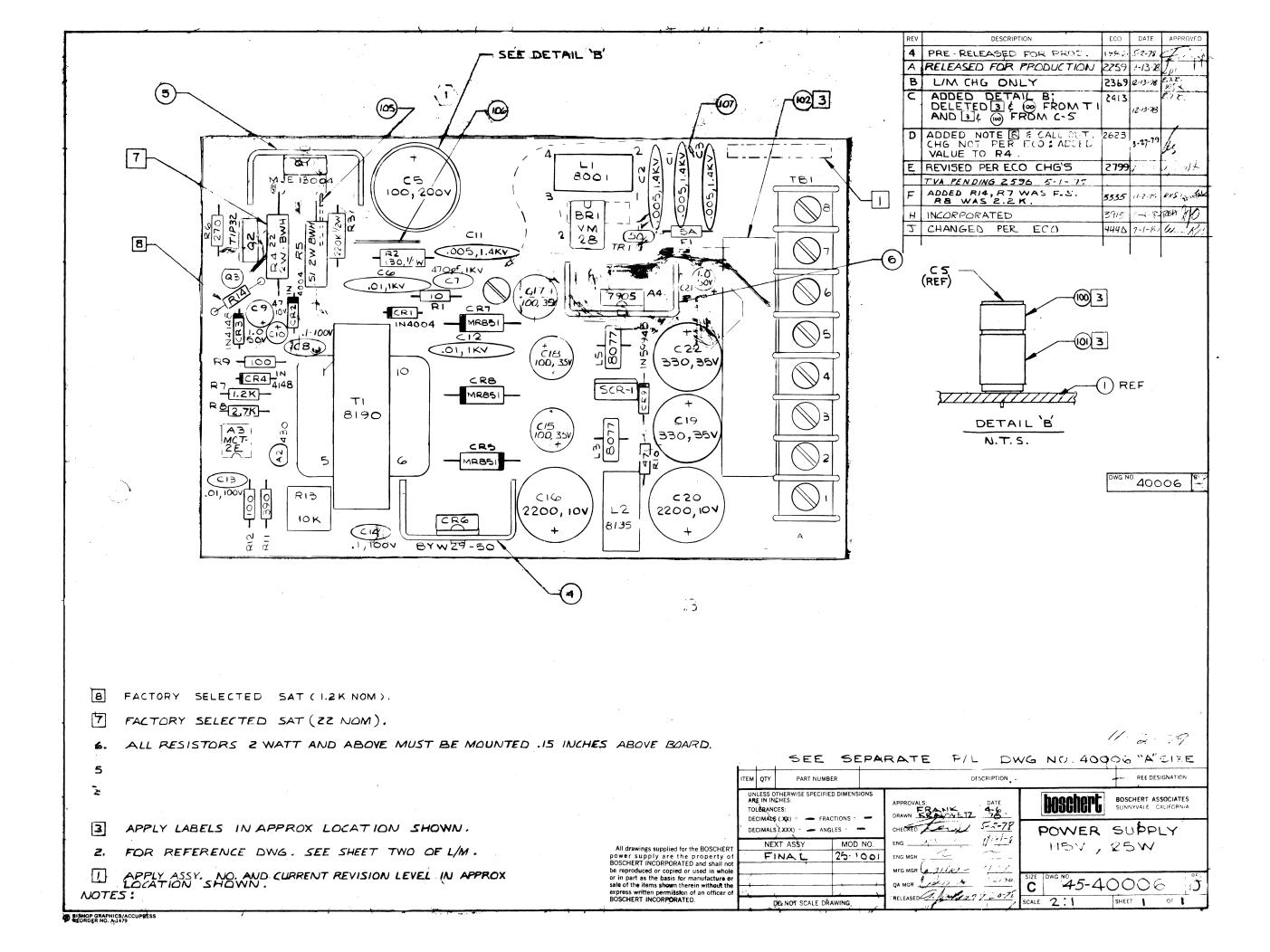
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCES ARE:	APPROVALS	DATE		bosebo		ERT INCORPORA	
FRACTIONS DECIMALS ANGLES	DRAWN Jonehus	11-6-79			301111	VALE, CALIFORI	
	CHECKED CHATZMAN	ZIN 80	PIN	$\sqrt{007}$	CHAR		
MATERIAL	ENG Howa ofthe	1-25-80		1 2		•	
0	TO COOLADOOR	2-6-80	^				
FINISH	OA MGR	2/7/80	SIZE	DRAWING NO		A –7	REV
	RELEASED SON	1-28-80	A		97-101	4 /	A
DO NOT SCALE DRAWING	C. Forego	178.60	SCALE	N/A		SHEET 1	OF Z

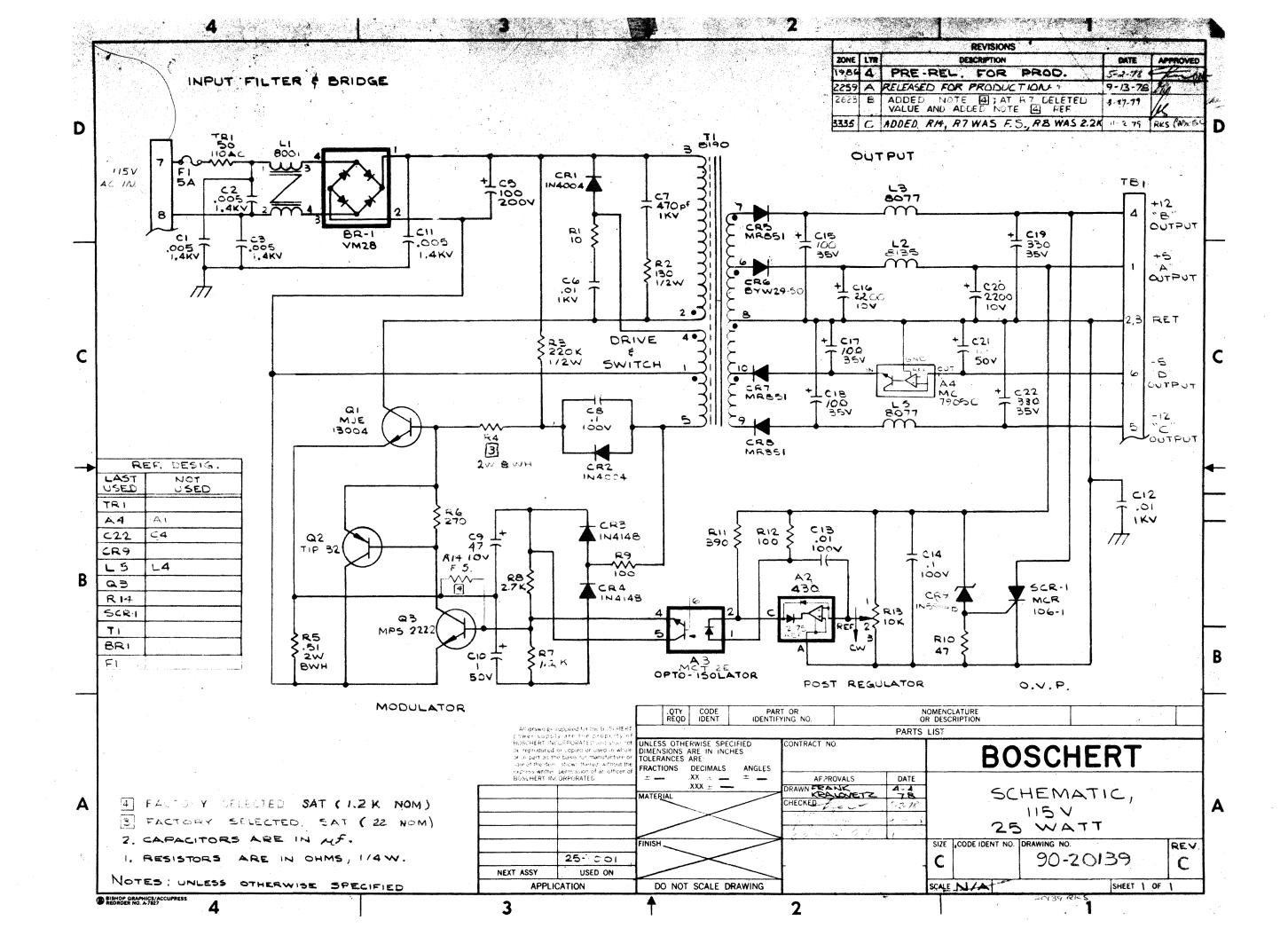
PIN ORIENTATION

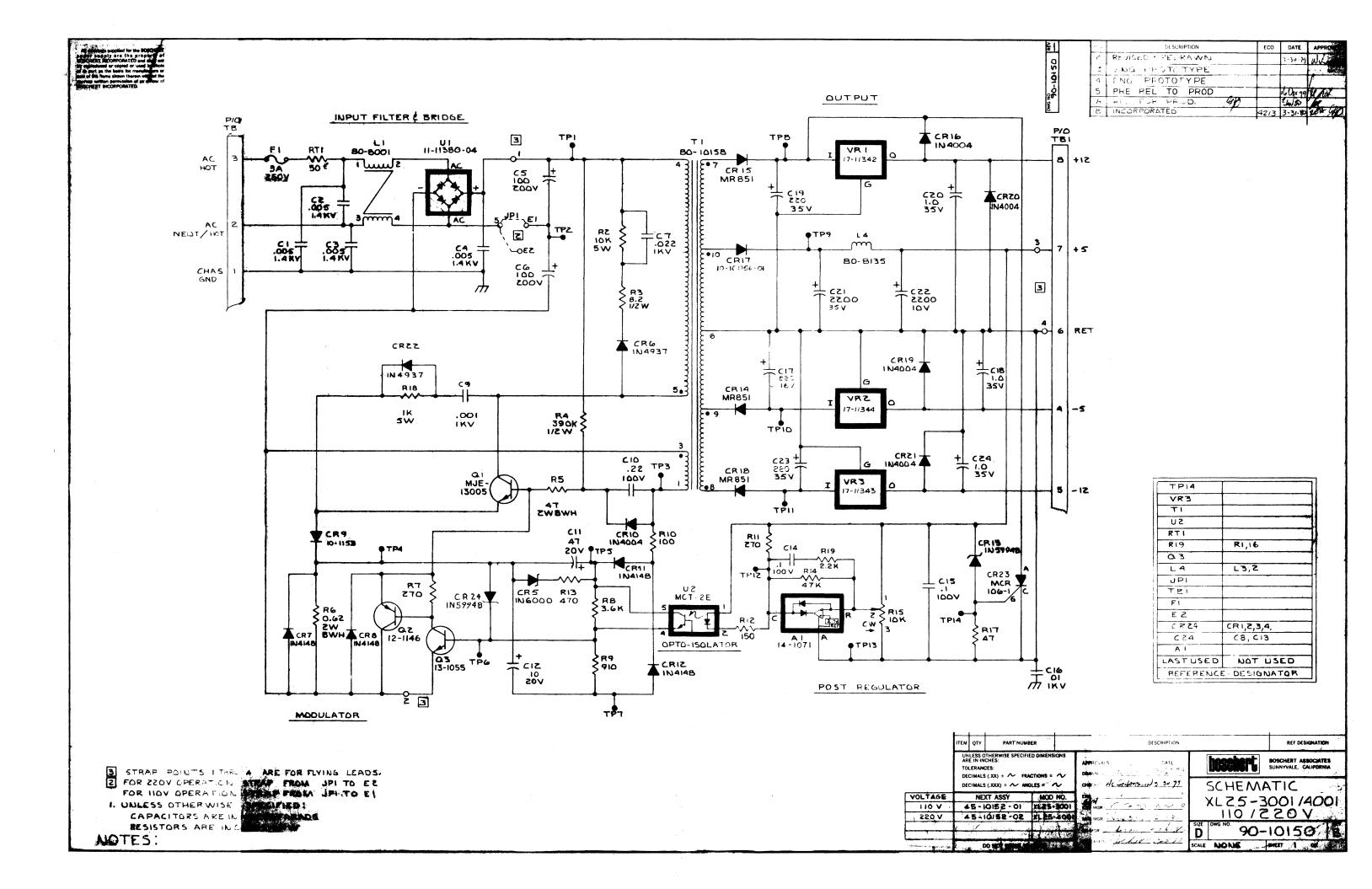


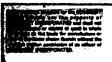
NOTE 1 - For 110V as shown; For 220V strap JP1 to E2 (applicable to the XL25-3001/4001 only).

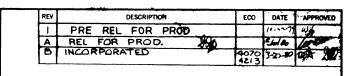
PIN	VOLTAGE	
2	INPUT A.C. NEUT/HOT	
3	INPUT A.C. HOT	
1	CHAS GND	MODEL NO. XL-25
4	OUTPUT	
5	OUTPUT12V	MFG. ASSEMBLY NO.
7	OUTPUT+5V	
. 8	OUTPUT+12V	
	OUTPUT	MATING
•	OUTPUT	CONNECTOR
	OUTPUT	
	OUTPUT	
6	RETURN	AUTOMATIC SHORT CIRCUIT RECOVERY
**************************************	NOT USED	AUTOMATIC SHORT CIRCUIT SHUT DOWN
		DWG. NO. 97-10147 REV A
		sht. 2 of 2

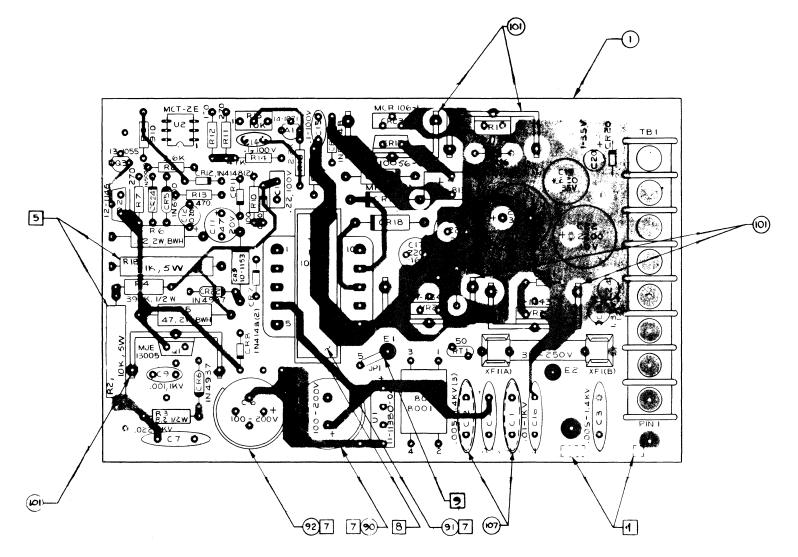












OPTION TABLE 1

ASSY P/N	MOD NO.	VOLT. RATING
45-10152-01	XL25-3001	110
45-10152-02	XL 25-4001	220

II UNLESS OTHERWISE SPECIFIED : RESISTORS ARE 1/4 W : DIODES ARE 1N4004 : CAPACITANCE IS IN 44.

10.

9 FOR -OI (110 V) STRAP JPI TO EI AS SHOWN; FOR -OZ (220V) STRAP JPI TO EZ

8 CORE MUST BE .15 ABOVE BOARD

APPLY LABELS IN APPROX POSITION SHOWN NOT COVERING THE COMPONENT IDENTIFICATION.

6

5 RZ & RIS MUST BE MOUNTED .30 IV. OFF THE BOARD.

4. ALL HEATSINKS MUST BE MOUNTED TO IN OFF THE BOARD.

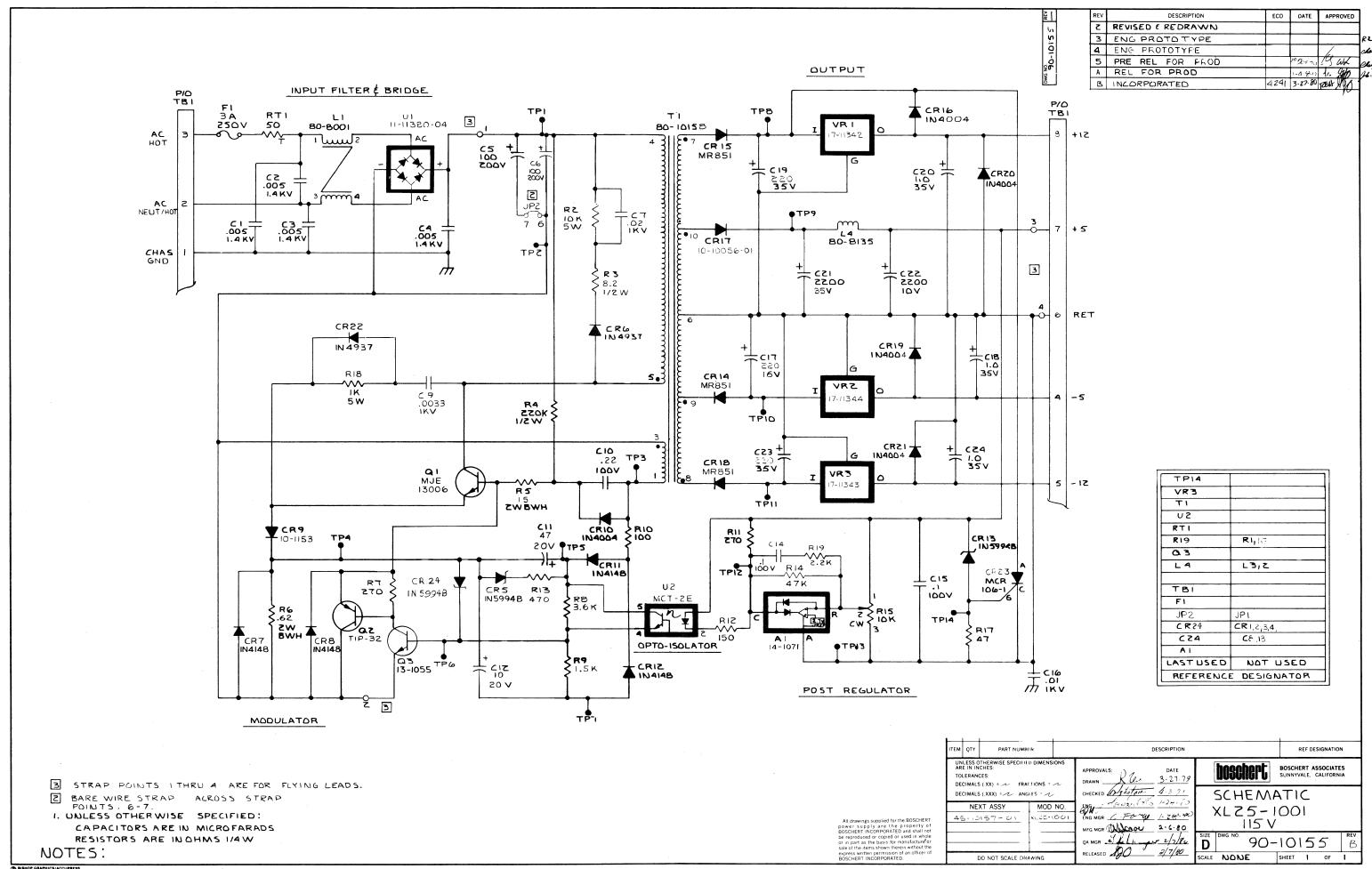
3. ALL RESISTORS TWO (2) WATTS AND ABOVE MUST BE MOUNTED .15 IN OFF THE BOARD UNLESS OTHERWISE SPECIFIED.

2. FOR REFERENCE DWG SEE LIST OF MATERIAL.

MARK ASSY DENT NO & CURRENT REV LEVEL IN APPROX LOCATION SHOWN PER OPTION TABLE.

NOTES:

EM QTY PAR	NUMBER	DESCRIPTION		REF DESIGNATION
UNLESS OTHERWISE S ARE IN INCHES: TOLERANCES: DECIMALS (.XX) ±	PECIFIED DIMENSIO	APPROVALS: DATE DRAWN Classes W. J.	beschert	BOSCHERT ASSOCIATES SUNNYVALE, CALIFORNIA
DECIMALS (.XXX) ±	ANGLES ±	CHECKED SOMETHING 19 OF 79	ASSE	MBLY
NEXT ASSY	MOD N	ENG WOL JULY	XL25 - 300	1/4001
FINAL	XL25-3	DOI ENG MAR LESCOTE of Por 1-25-0		
FINAL	XL25-4	001 BANG (FOW - 28-50	110/2	20 V
· · · · · · · · · · · · · · · · · · ·		WHQ498-11;89	SIZE DWG NO.	6EV
	\	OA MOR Site Library Wille	D 45-1015	2-01/-02
DO NOT SO	LE DRANGING	MELEASED JACKS 125-80	SCALE 2:1	SHEET 1 OF A



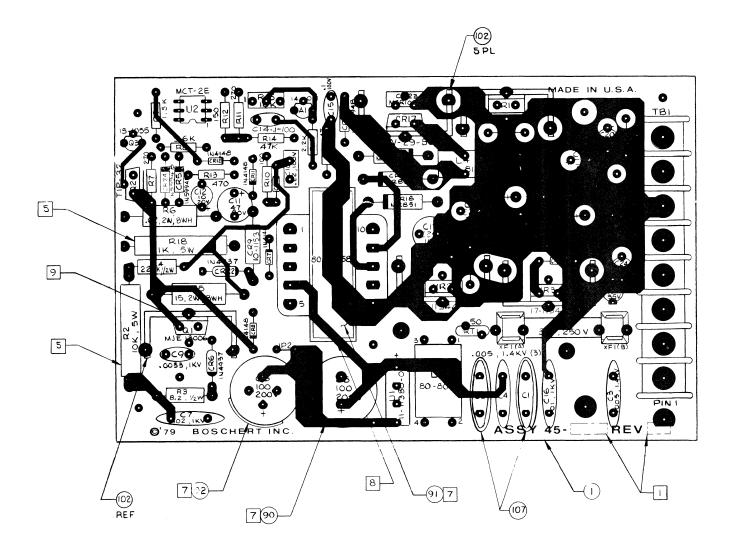


TABLE A

ASSY NO	MODEL NO	VOLT RATING
45-10157 - 01	XL25-1001R	115 V

10. UNLESS OTHERWISE SPECIFIED: RESISTORS ARE 1/4 W; DIODES ARE IN 4004; CAPACITANCE IS IN at.

9 APPLY A THIN COAT OF HEATSINK COMPOUND (P/N 74-7765) BEFORE ASSY.

8 CORE MUST BE .IS ABOVE BOARD.

7 APPLY LABELS IN APPROX POSITION SHOWN NOT COVERING THE COMPONENT IDENTIFICATION.

5 R2 & RIB MUST BE MOUNTED . BO OFF BOARD.

4. ALL HEATSINKS MUST BE MOUNTED .10 IN. OFF THE BOARD.

3. ALL RESISTORS TWO (2) WATTS AND ABOVE MUST BE MOUNTED .15 IN. OFF THE LIMED UNLESS THERWILE SPECIFIED.

2, FOR REFERENCE DWG SEE LIST OF MATERIAL.

I MARK ASSY IDENT NO & CURRENT REV LEVEL IN APPROX LOCATION SHOWN PER TABLE A

NOTE 5:

	ITEM	QTY	PART NUMB	IFR			DESCRIPTION			REF DESI	GNATION
	AR TO	E IN IN LERANO	CES:	CTIONS +	ions /	APPROVALS:	DATE 10_16-79		boschert	BOSCHERT ASS	
All drawings supplied for the BOSCHERT		CIMALS NE		MOD		ENG GENERALLE ENGLISHED ENGLISHED TO THE ENGLISHED ENGLISHED TO THE ENGLISHED TH	10 24 1 - 20 1-18-12)		ASSEI XL25 - 100		
power supply are the property of BOSCHERT INCORPORATED and shall not be reproduced or copied or used in whole or in part as the basis for manufacture or sale of the items shown therein without the express written permission of an officer of BOSCHERT INCORPORATED.			DO NOT SCALE DR	AWING		MFG MGR DALCOC QA MGR JA	2.6.80	SIZE D SCALE	DWG NO. 45-1015	7	OF I